

Atty. Docket No. 157622-0022  
Express Mail Label No. EL696986399US

UNITED STATES PATENT APPLICATION

FOR

**SYSTEM AND METHOD OF UTILIZING  
POSITIONING RECEIVER HARDWARE FOR  
NETWORK-BASED TRANSCEIVER APPLICATIONS**

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## CROSS REFERENCE TO RELATED APPLICATIONS

This application is related to and claims priority from provisional application serial number 60/429,866, entitled "System and Method of Utilizing Positioning Receiver Hardware for Network-Based Transceiver Applications," 5 filed on November 27, 2002.

## BACKGROUND OF THE INVENTION

### 1. Field of the Invention

The present invention relates in general to data networks, and more particularly to a networked navigation-enabled system and positioning receiver 10 hardware for network-based transceiver applications.

### 2. Related Art

While the development of navigation technology is a rapidly growing industry, the value of being able to remotely obtain precise position information has long been recognized. Numerous navigation applications have recently been 15 recognized and systems relating thereto developed, with the result being that navigation technology has found its way into cars, boats, planes, construction equipment, farm machinery and cellular phones.

One of the most well developed navigation system is the global positioning system (GPS). In fact, GPS technology has matured to the point that virtually everyone, including scientists, sportsmen, farmers, soldiers, pilots, surveyors, hikers, delivery drivers, sailors, dispatchers, lumberjacks and fire-fighters can benefit from it. Various applications in which GPS has been used include location, navigation, tracking, mapping and timing.

Recently, the idea of network-assisted GPS systems has been introduced. However, these systems have only scratched the surface of the applications that are possible by combining GPS technology with the full functionality of network communications. Thus, there is a need in the art for a fully-networked navigation-enabled system. In addition to the numerous possible applications in which a fully-networked navigation-enabled system may be used, there is a further need in the art for positioning receiver hardware for network-based transceiver applications that make use of navigation technology, such as GPS.

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BRIEF SUMMARY OF THE INVENTION

A network, network device and method is disclosed. In one embodiment, a method includes transmitting communication signals from a first network node to a second network node, where the first and second network nodes comprise a network and each include a receiver portion and a transponder portion. The method further comprises receiving, by the first and second network nodes, position signals from a plurality of navigation beacons, and generating transmitter codes for the transponder portions using local code generators of the receiver portions.

10 Other embodiments are disclosed herein.

BRIEF DESCRIPTION OF THE DRAWINGS

Figure 1 is a schematic diagram of one embodiment of a transponder consistent with the principles of the invention.

Figure 2a is a schematic diagram of another embodiment of a transponder  
5 consistent with the principles of the invention.

Figure 2b is a schematic diagram of yet another embodiment of a transponder consistent with the principles of the invention.

Figure 3a illustrates the simplified configuration and node interaction of a network, according to one embodiment.

10 Figure 3b illustrates the configuration and node interaction of a network, according to another embodiment.

Figure 4a depicts one embodiment of the functionality of a network consistent with the invention.

Figure 4b depicts another embodiment of the functionality of a network  
15 consistent with the invention.

## DETAILED DESCRIPTION OF THE EMBODIMENTS

One aspect of the invention is to provide a network of data devices having data representations of connectivity, network node position, and/or position topologies. In one embodiment, the data devices are nodes of the network and

5 have the ability to share data with each other, as well as data from other sources connected to any of the network nodes. In yet another embodiment, a database of network topology is distributed among the networked data devices such that changes in the network state would be incorporated into that distributed topological data representation.

10 In one embodiment, network nodes are addressable in terms of device identification, connectivity topology and/or according to spatial parameters. Such spatial parameters may include position, velocity and/or other parameters associated with dynamic or static behavior. Addressing of the network nodes (which in one embodiment are data devices) would enable communications of

15 data and control information to only the desired recipients separable by one or all of the parameters defining the network topology characteristic.

One application of having spatially addressable network nodes is the ability to define a method for secure communication. In one embodiment, communications between network nodes may be encrypted based upon

positional data, device ID or motional parameters increasing the difficulty of deriving information from the message. In another embodiment, positioning signal qualities that enable a PVT (position, velocity, time) calculation to be made may be used in an encryption scheme. For example, it is possible to reuse 5 positioning signals (or portions thereof) to generate a new signal that can be correlated and therefore demodulated only by the recipient in the intended position space. This spatial-correlation quality would enable a reciprocal communication link to be established between one or more qualified spaces, thereby diminishing the ability of the communication to be compromised by 10 those outside the intended qualified spaces. Similarly, in another embodiment, the timing precision and synchronicity enjoyed by the network may be used in traditional secure communications techniques, such as direct sequence spread spectrum ("DSSS") or time hop spread spectrum.

Another aspect of the invention is to provide a self-configuring network. 15 In one embodiment, the network can self-configure to add or remove network nodes on a connectivity basis. New nodes (e.g., data devices) not previously on the network can join by protocol sharing with the network after the connection is made, according to one embodiment. Similarly, the network may change the status of a node no longer connected to an inactive status. In yet another

embodiment, the last known parameters for a removed node are distributively stored after criteria for disconnection is met.

Another aspect of the invention is to provide a data positioning device capable of operating as a node in a network of the present invention. In one 5 embodiment, the data device is a radio frequency device having integrated into it means for wireless communication, remote control and/or encrypted communication for use in a network as described above. In yet another embodiment, the data device is a portable low cost and low power device capable of receiving signals from navigation beacon transmitters and sensing 10 environmental characteristics for calculation of position and dynamic behavior. Moreover, the data device may further be capable of relaying that position and motion information to another device via a communications link, which may or may not be part of the network of data devices.

As mentioned above, one aspect of the invention is to combine 15 transponder and receiver functionality. In the case of a direct sequence spread spectrum ("DSSS") positioning receiver, given the fact that DSSS is based upon the reciprocal properties of modulo-2 addition, much of the receiver hardware may be shared with transmitter applications. Since DSSS modulation and demodulation are reciprocal processes, the bi-phase modulation of the carrier

wave in a transmitter consists of multiplying the pseudo-random noise (PRN) sequence with the carrier wave. If we can multiply that bi-phase modulated waveform with a synchronized, identical PRN sequence, it is possible to duplicate the original carrier wave. Conversely, if we multiply the bi-phase modulated waveform with the original carrier wave, we get the original PRN sequence as the result. These reciprocity characteristics lead us to the conclusion that the locally-generated code can be used to "correlate" with the received bi-phase modulated code and yield a resultant carrier wave representation. So, in effect, the local code generator in a receiver is identical to the associated transmitter code generator and therefore may be used as a transmitter code generator itself.

In general, the sensitivity of a receiver is inversely proportional to the bandwidth of that receiver, reaching a sensitivity maximum at the matched filter bandwidth and then degrading with increasingly narrow filters. As a result, the time it takes to find a signal is inversely proportional to the signal level and proportional to the frequency uncertainty which sets the number of searches for the signal in the frequency domain. Providing precise signal frequency assistance and signal phase assistance reduces the number of signal search operations necessary to find the signal, which promotes prompt position determination in signal-impaired conditions where the mandated narrow

bandwidths result in longer detection times. Thus, another aspect of the invention is to provide signal-based assistance information to increase the ability of the data positioning device to find attenuated positioning signals using the characteristics of the data signal itself.

5        In one embodiment, the network design enables the network communications signals to be used as substitutes for the network node position signals (e.g., navigation beacon signals). Substitution for the network node position signals requires that the communication signals have the same essential characteristics required of the network node position signals to facilitate position 10 calculations to be performed.

As will be described in greater detail below, where the network nodes are navigation beacons, the positioning network may require a system of equations derived from signals containing exact time, frequency and time-tagged beacon position information. The beacons of such a positioning network may transmit 15 signals with ranging properties that enable signal phase to be distinguished against universal precise time in the receiver, thereby yielding the range from each beacon to the receiver by factoring the speed of light into the calculus.

Since the proposed positioning network nodes are able to substitute for missing navigation beacon signals, they will also have the essential

characteristics for position calculation. Those signal characteristics may be representative of the assistance information that may be sent to the positioning receiver. In one embodiment, the communications signals are essentially a function of the navigation beacon signal, with the frequency having a known 5 relationship to the associated navigation beacon signal, and the communications signal PRN being a known function of the navigation beacon signal PRN. As a result, a positioning transponder could derive information for finding weak signals from a substantially stronger network communications link signal without requiring that the information be modulated onto the signal as data.

10 Thus, in one embodiment the positioning network may have unique intrinsic assistance information designed into its signals.

Referring now to the figures, Figure 1 illustrates one embodiment of an integrated transmitter/receiver unit consistent with the principles of the invention, which would enable continuous half-duplex communications and 15 positioning. In particular, Figure 1 depicts a twelve-channel positioning transponder 100 capable of receiving the position signals 48 by way of positioning antenna 2. In the embodiment of Figure 1, the positioning signals 48 are comprised of twelve CDMA (Code Division Multiple Access) positioning signals or CDMA communication channels.

In one embodiment, the position signals 48 may be comprised of one or more navigation beacon signals, as required for receiver position determination. While in one embodiment, the position signals 48 is generated from DSSS beacons, it should be appreciated that the position signals 48 may be provided by 5 any navigation system beacon capable of providing positioning information.

Once received, the position signals 48 may be amplified by amplifier 4, which in one embodiment is a low-noise amplifier. Thereafter, filter 6 may used to extract the components of the signal 48 at a desired frequency or range of frequencies. In one embodiment, filter 6 is a band-pass filter. It should further 10 be appreciated that other methods of signal processing may be used.

Regardless of the front-end signal processing employed, switch 8 is used to select the signal from either positioning antenna 2 or transponder antenna 10. When switch 8 is set to accept the position signals 48, RF downconverter 22 may be used to down-convert the signal 48. In one embodiment, RF downconverter is 15 an RF to IF converter. Next, an Analog to Digital Converter (ADC) 24 converts the position signals 48 from an analog signal to a digital signal before being provided to correlation/tracking circuit 50.

In one embodiment, correlation/tracking circuit 50 includes correlator/demodulator 30, PRN code generator 36 and tracking mechanism 34.

However, it should be appreciated that the functions of correlator/demodulator 30, PRN code generator 36 and tracking mechanism 34 may be performed by other circuits that do not comprise the correlation/tracking circuit 50. However, for convenience, the following discussion will assume the configuration shown 5 in Figure 1.

In one embodiment, tracking mechanism 34 performs clock generation, signal tracking and control functions. PRN code generator, based on input from tracking mechanism 34, provides correlator/demodulator 30 with PRN codes which are then used to correlate the received position signals 48 to the receiver 10 locally generated signal. In another embodiment, the position signals 48 from multiple beacons may be processed using parallel hardware channels 1-n, where n represents the total number of position signals 48 and the total number of corresponding hardware channels. In this embodiment, correlation/tracking circuit 50 would be comprised of PRN code generators 36<sub>1-n</sub> and tracking 15 mechanisms 34<sub>1-n</sub>.

Controller 46 is part of the CPU system required for most firmware-controlled hardware. DSP subsystem 44 functions as a CPU to control, calculate and schedule operations required for the transmitter and positioning receiver functions, while memory 42 is used by DSP subsystem 44 to execute instructions

required for transponder operation. However, it should obviously be appreciated that other processor configurations may be used. In addition, in the embodiment of Figure 1, controller 46 passes either maximal length codes or Gold codes to PRN code generator 36. Maximal length codes have the property 5 of not repeating until the  $2^{n-1}$  clock cycle passes and they have correlation properties defined by an up-slope and down-slope shape. Gold code, on the other hand, can be generated by modulo-two adding two maximal length codes together. While Gold code carries most of the characteristics of the maximal length code, it may have correlation responses besides the main one. These other 10 correlation responses should be smaller than the main response for the codes used in the GPS.

Continuing to refer to Figure 1, the transmission functionality of transponder 100 will now be described. Transponder antenna 10 may be used to transmit and receive communication signals 52 with other transponders. Given 15 that transponder antenna 10 is capable of two-way communication, T/R switch 12 is used to select between transmit and receive paths. Moreover, as with the position signals 48, a received communication signal 52 may be processed with amp 14 and filter 16. When receipt of the communication signals 52 is desired, switch 8 is set to accept the communication signals 52, which may then be 20 provided to RF downconverter 22. Sharing of RF downconverter 22 is enabled

using programmable frequency synthesizer 28, which may itself be shared for both receiving and transmitting functions. While it may be economically desirable to use RF downconverter 22 and ADC 24 for both the position signals 48 and the communication signals 52, it should be appreciated that separate 5 circuits for each may also be employed.

Transponder 100, like most positioning receivers which operate with DSSS signal techniques, uses correlators and PRN code tracking loops for signal de-spreading, synchronization and ultimately sensing the pseudo-range to the navigation system beacon being received on a given channel. The hardware to 10 perform these functions, which in the embodiment of Figure 1 is performed by correlation/tracking circuit 50, may be suitable for use in transmitter waveform generation. By making use of the reciprocity of the DSSS signal, it is possible to use the same tracking mechanism 34 and PRN code generator 36 for input to a complimentary code keying ("CCK") modulator to generate a transmission 15 signal that another receiver could in turn demodulate. Thus, in the embodiment of Figure 1 having parallel hardware channels 1-n, tracking mechanisms 34<sub>1-n</sub> and PRN code generators 34<sub>1-n</sub> provide tracking loop and code generation data to CCK modulators 36<sub>1-n</sub> for use in generating transmittable communication signals 52. The number of CCK modulators 36<sub>1-n</sub> may be added as needed up to a 20 maximum number defined by the number of different code generators and

tracking mechanisms available. As used herein, received communication signals 52 refers to those signals which are received by the transponder over transponder antenna 10 from other devices. Transmittable communication signals 52 refer to those signals that are generated by the transponder 100 and 5 which are sent to other devices via transponder antenna 10.

As mentioned above, transponder 100 may be used in a navigation beacon positioned network or, alternatively, in a non-navigation beacon relative-positioned network. In the embodiment of the beacon-positioned network, synchronization of the transmittable communication signals 52 to the incoming 10 position signals 48 may be employed. Tracking mechanisms 34<sub>1-n</sub> are used for those applications of the transponder 100 which require synchronization of the transmittable communication signals 52 to the incoming position signals 48 from the navigation system positioning beacon. In the embodiment where synchronization is required between the two systems, tracking loops of tracking 15 mechanisms 34<sub>1-n</sub> that have been in a steady-state tracking mode for positioning would then "flywheel" at the rate of change for positioning and be switched over for the transmission function and then switched back for positioning again. In other embodiments, independent communications devices between positioning and communications would enable simultaneous synchronized operations 20 between the network and positioning system.

However, in applications which use non-beacon relative positioning, a master transmitter in the network may be used to set the basic frequency and phase of the network and the non-master elements (e.g., other transponders 100) would then sample network signals for relative position and also switch to 5 transmit the communication signals 52 when required for data transmission, according to one embodiment.

In yet another embodiment, these transmittable communication signals 52 may also double as ranging signals for the other receivers/transponders.

Ranging signals employed by positioning or navigation systems may use several 10 different methods of time tagging the signal for propagation time determination from a transmitter to a positioning receiver. The time tagging enables the receiver to determine the transmit time which is modulated onto the signal by the transmitter and compare the demodulated time tag to the local time reference to determine the time of propagation for the signal. This determined 15 propagation time, scaled by the speed of light, may then be used to determine the range of the transmitter to the receiver. In such a positioning network, the system of known navigation beacon positions at their transmitted signal time tag yields a system of equations involving time, the transmitter positions at time, and the range of the receiver to those transmitter known positions. This system 20 of equations usually has four variables and at least four equations which then

can be solved for either a determined or over-determined solution. The variables may include x, y, and z position data, as well as time. By way of providing a non-limiting example, one embodiment of a system of equations with four navigation beacons would be as follows:

$$\begin{aligned}\sigma_{1r}^2 &= (x_1 - x_r)^2 + (y_1 - y_r)^2 + (z_1 - z_r)^2 + (c \times t_{\text{universal}})^2 \\ \sigma_{2r}^2 &= (x_2 - x_r)^2 + (y_2 - y_r)^2 + (z_2 - z_r)^2 + (c \times t_{\text{universal}})^2 \\ \sigma_{3r}^2 &= (x_3 - x_r)^2 + (y_3 - y_r)^2 + (z_3 - z_r)^2 + (c \times t_{\text{universal}})^2 \\ &\vdots \\ \sigma_{nr}^2 &= (x_n - x_r)^2 + (y_n - y_r)^2 + (z_n - z_r)^2 + (c \times t_{\text{universal}})^2\end{aligned}$$

5 where :

$\sigma_{nr}$  = pseudorange of beacon  $n$  to receiver based on code phase measurement

$x_n$  = beacon  $n$  coordinate, x axis

$y_n$  = beacon  $n$  coordinate, y axis

$z_n$  = beacon  $n$  coordinate, z axis

$c$  = speed of light

$t_{\text{localuniversal}}$  = best fit solved for universal time based upon best position fix result

This system of equations may be used to enable a best-fit solution of receiver position and time based upon received signal phase from four navigation beacon signals.

10 In navigation systems where the positioning receiver can calculate position, it may be preferable for there to be some receiver knowledge of each of the navigation beacon's position and knowledge of the "universal time" of the transmitted navigation beacon signals. In cases where those navigation beacon

positions are not known, it may be desirable to convey the navigation beacon transmitter time-position to the receiver. To that end, in one embodiment such data may be encoded onto the navigation beacon transmitter signal, thereby enabling the receiver to demodulate the information and use it in the position 5 calculation.

In one embodiment, a navigation beacon signal construct that adheres to the description above may utilize a CCK BPSK ("Bi-Phase Shift Keying") signal that has temporal properties useful for ranging measurements, while also enabling data modulation. In one embodiment, such a CCK BPSK signal may be 10 described mathematically as:

$$S_{beaconID}(t, prncode_{ID}(t), message_{ID}(t)) = \cos(\omega_{c, ID} * t + \frac{\pi}{2} * message_{ID}(t) * prncode_{ID})$$

where the signal  $S_{beaconID}(t, prncode_{ID}(t), message_{ID}(t))$  consists of :  
a function of time,  $t$ ;  
the PRN sequence, which may also function as the beacon ID,  $prncode_{ID} \in \{-1,1\}$ ;  
a carrier wave,  $\cos(\omega_{c, ID} * t)$  with the angular carrier frequency associated with that particular beacon and therefore the associated beacon ID; and  
a data message,  $message_{ID}(t) \in \{-1,1\}$ , which represents a time mark definition and the time - position function of the beacon that sent it.

Given that the PRN code has a cycle associated with the universal time used by the navigation beacon network, and that the data message contains

information about the universal time and positions of the beacons, in one embodiment the receiver would be able to extract sufficient information to calculate its own position.

The data message and the PRN code may both have binary properties and

5 have values for this argument of "1" or "-1". It is also possible to alter that PRN through a process known as CCK, which is a modulation wherein a PRN sequence is inverted when the message bit is a "logical 1" and not inverted when it is a "logical 0". In order to mathematically represent this process, the data message and the PRN are considered as bi-polar signals; namely +1 and -1 for

10 logical 1 and logical 0, respectively. As the BPSK process involves inversion of the signal carrier for a "logical 0" and a pass-through for a logical 1, a multiplication of the carrier wave by the bi-polar representation will produce a BPSK modulation.

A local representation of the received CCK BPSK code may be produced

15 in the receiver in order to enable detection of the transmitted signal and to demodulate any message that would be encoded upon it. In one embodiment, the receiver can synchronize the local PRN to the PRN modulation riding on the transmitted signal. This is accomplished through the correlation properties of the PRN, which are at a maximum (a normalized +1) for correlation of like-PRN

sequences which have the same “polarity sense” and which are in phase with each other. Similarly, the correlation properties of the PRN are at a negative maximum (a normalized -1) for correlation of like-PRN sequences which have the opposite “polarity sense” and which are in phase with each other.

5        In one embodiment, each PRN sequence has a length after which the PRN repeats itself. This length, which may be referred to as the “PRN cyclicity”, may also help to define the correlation response to a PRN sequence that is unsynchronized with respect to a PRN replica that is locally generated. For maximal length PRN sequences, the unsynchronized response may reduce to the

10      normalized level of the value defined as “polarity sense\*(-1/(code length),” where the polarity sense is +1 or -1 when the two sequences are unsynchronized by one or more PRN clock periods.

As a result, the correlation response between the received PRN and the locally-generated PRN can be used as a measurement of synchronization by

15      having a triangular response in delay space with the maximum correlation excursion occurring when the two are synchronized. To that end, in one embodiment, PRN synchronization uses these principles with a hypothesis-test loop where the locally-generated PRN is multiplied with the receive signal and the response is measured. The phase of the locally-generated PRN may be

adjusted and the multiplication measurement response evaluated until an acceptable multiplication response level is attained. As the response is a triangular function in synchronization space, any loss of synchronization will result in a loss of correlation from the multiplication process and can be used to 5 correct the PRN clock rate and/or PRN phase to maintain the maximum level of synchronization.

In order to detect advanced or retarded local PRN phase, two or more multipliers may be used with a fraction of a PRN clock cycle signal delay between them (e.g.,  $\frac{1}{2}$  or  $\frac{1}{4}$  the PRN clock cycle). In such a case, when the local 10 code was advanced the advanced or "early" correlator would increase in response, and the "late" correlator would decrease in response. Conversely, when the local code was retarded from the receive signal the "early" correlator would decrease in response, and the "late" correlator would increase in response. Thus, an "early response-late response" term may be used to provide 15 a feedback control signal which could be used to retard or advance the locally generated PRN replica rate towards maintaining receiver synchronization. This delay lock loop (DLL) technique may operate via controlling the locally generated PRN rate such that the received signal PRN phase is maintained between the early and late correlators, where an equal response for both may 20 indicate the best possible synchronization.

Through the properties of a bi-polar multiply, it is also possible to demodulate the data from the received signal by multiplying it with a locally generated replica signal consisting of locally-reproduced carrier wave multiplied with PRN. In another embodiment, the above described early and/or late 5 correlator outputs may also be used for this second purpose. The local code phase, which is known by virtue of being synchronized with the received signal through the above described DLL process, may also provide the code phase measurement necessary for ranging from the receiver to each of the respective navigation beacons.

10 In another embodiment, there may be an error rate above which the desired position information can no longer be extracted. Given that the desired position information is used to calculate position, this may cause a positioning sensitivity limitation. The code phase may be measured with an almost arbitrary level of sensitivity given sufficient local oscillator stability, sufficiently narrow 15 bandwidths, sufficiently long integration times, etc. It is possible to "assist" such a positioning receiver with information that would normally be sent with the data message via some other communications link. For example, it is possible to send over a separate communications link the range of code phases expected at the receiver position, the frequency of the navigation beacon signal, the 20 identifications (and PRN sequences therefore) of the navigation beacons, etc.

This would be useful in reducing the amount of time it takes a receiver to measure the code phase since the narrow bandwidths necessary for weak navigation signal detection require a proportionately longer measurement period. Furthermore, because the bandwidths are narrow, the number of search 5 attempts is increased with an indeterminate frequency of the signal.

While the information needed to perform these assistance functions may be sent as explicit data over the communications link, in one embodiment this information is implicit in the communication signal in question. That is, the PRN 10 code cycle used by a transponder signal is related to the navigation beacon signal PRN code cycle such that a receiver of the communications signal would have the PRN code timing of the "companion" navigation signal it is associated with. This is a form of code phase assistance that is implicit in the signal, rather than an explicit data representation of that code phase assistance. Likewise, the transponder communications signal frequency is deterministically related to the 15 associated "companion" navigation signal. Therefore, timing information would be part of the communications data message and would identify the universal time associated with the next code phase cycle commencement.

After controller 46 provides transmission data 62 to the CCK modulators 36<sub>1-n</sub>, PRN code data 56 and correlator data 54 for a given channel n may be

combined and summed by summers 38 for channels 1-n to provide a CCK signal 64. Thereafter, the CCK signal 64 may be up-converted in order to change the basic CCK signal 64 frequency to the desired frequency of the transmittable communication signals 52. In the embodiment of Figure 1, this is accomplished 5 with the programmable frequency synthesizer 28 which provides the required frequency to up-convert to the frequency of the transmittable communication signals 52. The use of the programmable frequency synthesizer 28 for both the up-converter 26 and RF down-converter 22 follows from the fact that the up- conversion and down-conversion processes are reciprocal processes. A 10 transmittable communication signal 52 may be processed by amp 66 and filer 68 before being transmitted via transponder antenna 10.

Moreover, transponder 100 has been shown with two separate antennae because it is likely that any transmitted communication signals 52 would be much larger in amplitude than the position signals 48, which may interfere with 15 positioning receiver operation. Also, some navigation beacon systems are "radio transmission" protected to prevent emissions, intentional or unintentional, from interfering with the navigation signals which may be small in amplitude. Because of this protection against emissions, it may be desirable to choose other sanctioned frequencies for the communications function of the positioning 20 network transponders. However, absent these factors, it should be appreciated

that the position signals 48 and communication signals 52 may be received on the same antenna.

Referring now to Figure 2a, another embodiment of transmitter 100 is depicted showing additional detail of correlation/tracking circuit 50. Moreover, 5 in this single-data-channel embodiment, up-converter 26 is replaced with RF mixer 70, which operates in conjunction with XOR gate 75 and universal synchronous transmitter 80, as depicted in Figure 2a. In particular, PRN code data 56 is provided to universal synchronous transmitter 80 and correlator data 54 is provided to XOR gate 75. Thereafter, using the output of the universal 10 synchronous transmitter 80 and the correlator data 54, the XOR gate 75 generates the CCK signal 64, according to one embodiment. In this embodiment, rather than use up-converter 26, the transmittable communication signal 52 is generated by direct conversion using binary phase shift keying (BPSK) in the RF mixer 70 via multiplication of the CCK signal 64 (provided by the XOR gate 75) with the 15 local oscillator generated by the programmable frequency synthesizer 28.

Figure 2b depicts a multi-channel embodiment of transponder 100 that includes a complex mixer for single-sideband up-conversion of the I and Q summed BPSK-CCK signals to a BPSK-CCK signal at the RF carrier frequency.

In this embodiment, communication bandwidth is maximized by having a communications channel associated with and sharing hardware and software.

It should be appreciated that one or more of the components comprising transponder 100 may be implemented as integrated circuit(s), firmware and even 5 software. Moreover, one or more of the components comprising transponder 100 may be implemented by the chipsets marketed under the tradename of u-Nav Microelectronics Corporation, such as those chipsets having model number designations of uN8021B and uN8031B. The principles of the invention may also be implemented by modifying the chipsets marketed under the tradename of 10 u-Nav Microelectronics Corporation, such as those chipsets having model number designations of uN8021B and uN8031B.

Referring now to Figure 3a, in which a network consistent with the principles of the invention is depicted. In particular, positioning network 300 is shown comprised of a number of transponders 100 that are able to communicate 15 with each other via communication signals 52, where such signals may include data, position information, network topology and/or commands. In addition, positioning network 300 (and each node therein) can communicate with other networks 320. Moreover, in the embodiment of Figure 3a, positioning network

300 receives position signals  $48_{1-n}$  from navigation beacons  $310_{1-n}$ , which in one embodiment are DSSS satellite systems.

As mentioned previously, one aspect of the invention is to have network nodes that are addressable by physical characteristics. Thus, in the embodiment 5 of Figure 3a, the nodes of positioning network 300 are the transponders 100, which may each have knowledge of the network physical topology via communication signals 52. In one embodiment, the node-level knowledge is maintained on a real-time basis. In another embodiment, the addressability of the nodes (e.g., transponders 100) is based upon physical quantities of either 10 relative or absolute location, velocity or any combination thereof. Node identification can also be incorporated into the addressing scheme where each transponder 100 has its own device ID.

As previously mentioned, one application of having spatially addressable network nodes is the ability to define a method for secure communication. To 15 that end, in one embodiment, communication signals 48 may be secure communications, encrypted with position authentication. In another embodiment, communication signals 48 may be spatially encoded accessible only at a desired position. A spatially encoded signal could be demodulated only by a receiver at the proper position. In one embodiment, this scheme could be

utilized in "range-gated" communications in LPD applications and for spatially selectable communications and control functions. Similarly, other signal qualities may be used to enable a PVT calculation to be made and used in an encryption scheme. Any combination of PVT outputs, or even raw receiver data 5 (e.g., code phase, etc.) could be used as a means of selecting intended recipients.

In another embodiment of spatial addressing, the positioning network could provide the necessary components to the desired signal from separate sources. By way of example, in a three-source case, if necessary information were available only if signal X at range of WX were combined with signal Y at 10 range WY and with signal Z at range WZ, only one point would provide the ability to receive and combine the three signals with the proper relationships for extraction of the desired information.

In yet another embodiment, positioning network 300 is a self-configuring network. In particular, new transponders 100 can join the positioning network 15 by protocol sharing, according to one embodiment. Due to the self-configuration properties and distributed storage capabilities of positioning network 300, data relay becomes possible where network nodes (e.g., transponders 100) that cannot communicate directly, can pass data to each other through interstitial nodes. By use of this data-relay function, any node that receives position signals 48 of

suitable quality for position fixing can pass the necessary data to the rest of the network nodes that may be located in weak-position-signal conditions. Thus, one aspect of the present invention is to provide location capability for transponders 100 in weak-signal environments.

5        With respect to populating the network with network topology data, relative positioning of network nodes is possible where nodes calculate position data from communication signals 52 that are received. The transmitting nodes on the network would have position knowledge, either absolute or relative, and the triangulation techniques used commonly in navigation positioning receivers

10      would operate in the same way on the network transponder frequencies.

Moreover, since all of nodes will have position signals 48 or communication signals 52 or both, it is possible for all nodes to be co-synchronized to each other. Thus, if position signals 48 are available to any node of the network, network synchronization to the navigation beacon 310 system is possible.

15      In yet another embodiment, VLBI (Very Long Baseline Interferometry) measurements are facilitated without the need for additional hardware. VLBI techniques are typically used in some high-accuracy positioning systems and involve the combination of the positioning data from two or more receivers for the position calculation process. The required data may be passed over the

positioning network 300 between the nodes, which in the embodiment of Figure 3a are the transponders 100. The use of positioning system time base may be used as a frequency reference for any systems combining other functions requiring precise frequency reference.

- 5        As previously mentioned, one aspect of the invention relates to network assisted positioning in which a communications link provides information to assist a positioning receiver in acquiring the navigation signal. In this fashion, sensitivity and acquisition times of positioning receivers can be improved. In one embodiment, this is accomplished by having the communications signals carry implicitly in their characteristics (e.g., frequency, ranging code phase, etc.) a known relationship to the associated navigation signal desired for acquisition.
- 10      It is noteworthy that using these signal characteristics in this fashion enables position assistance functionality to be performed without using data representations of frequencies, code phases, and beacon PRN sequences.
- 15      Relative positioning is also possible using triangulation techniques and communications signals for ranging measurements. In another embodiment, a communications channel may augment the navigation beacon 310 position signals 48.

In the embodiment of Figure 3b, no navigation beacons 310 are used. Rather, transponders 100 exchange positioning information as was described above with respect to Figure 3a, but on a relative basis. Similarly, network topology data is maintained on a relative basis. Thus, the positioning network 5 300 has knowledge of relative positions between the communicating network nodes (position transponders 100) by mutual triangulation of the system of signals passing between the nodes (e.g., communication signals 52). If at least one absolute position is known, the entire network is position self-aware.

Figure 4a depicts a simplified functionality diagram of a positioning 10 network 300 consistent with the principles of the invention. Figure 4b is a more detailed functionality diagram of positioning network 300 showing numerous embodiments of the functions enabled by the invention.

While certain exemplary embodiments have been described and shown in the accompanying drawings, it is to be understood that such embodiments are 15 merely illustrative of and not restrictive on the broad invention, and that this invention not be limited to the specific constructions and arrangements shown and described, since various other modifications may occur to those ordinarily skilled in the art.